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# 전자장 데이터 기반의 전류분포 모델링: 극소 다이폴 모델링





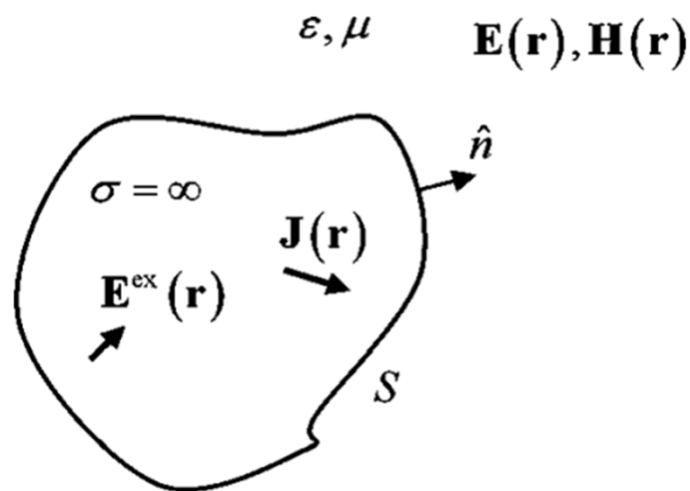
## Outline

- 01** 전류분포 모델링: 극소다이폴 모델링
- 02** 극소다이폴 모델의 검증
- 03** 전류분포 모델링의 응용

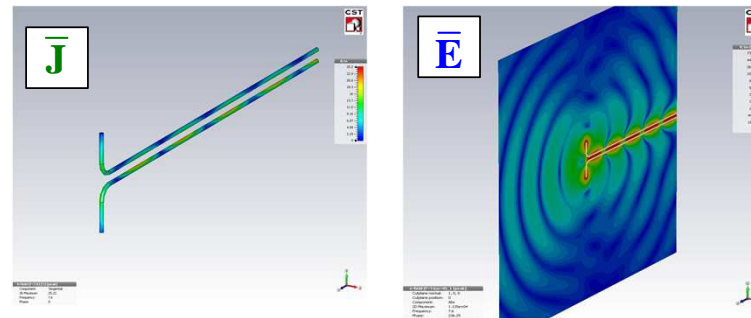
# 전류분포 모델링: 극소다이폴 모델링

## 전류분포와 Free-space Green's function

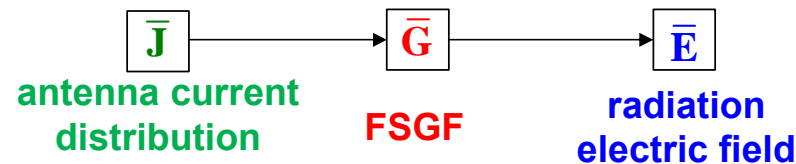
- how to calculate radiated field from current distribution?
- through **Free-Space Green's Function (FSGF)**
- $\mathbf{E}^{\text{ex}}$  (incident field)  $\rightarrow \mathbf{J}$  (induced current)  $\rightarrow \mathbf{E}, \mathbf{H}$  (radiated field)



transmitting mode analysis using FSGF



$$\bar{\mathbf{E}}(\mathbf{r}) = \int_S ds' \bar{\mathbf{G}}(\mathbf{r}, \mathbf{r}') \cdot \bar{\mathbf{J}}(\mathbf{r}')$$

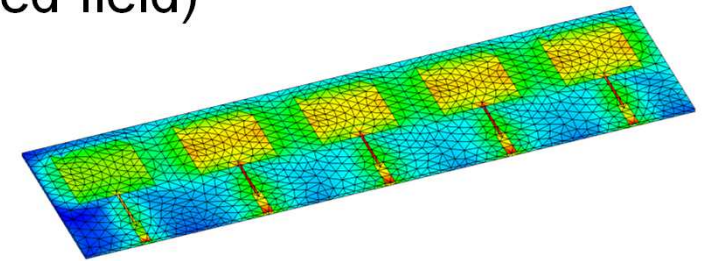




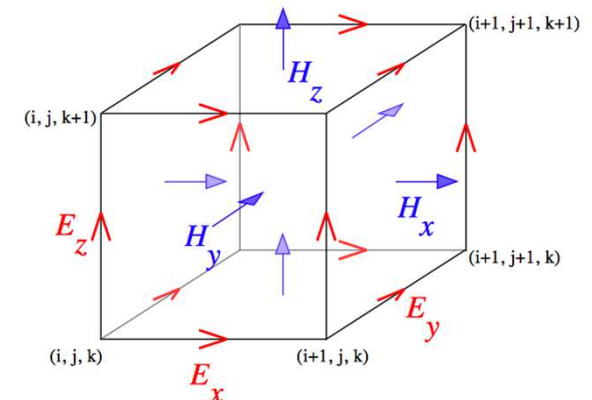
# 전류분포 모델링: 극소다이폴 모델링

## 전류분포와 Free-space Green's function

- $\mathbf{E}^{\text{inc}}$  (incident field)  $\rightarrow \mathbf{J}$  (induced current)  $\rightarrow \mathbf{E}, \mathbf{H}$  (radiated field)
- the most time, memory, cost-consuming process
- 전류 분포는 어떻게 구할까?
  - analytical: solving electromagnetic problems
  - numerical: full-wave methods (MoM, FEM, FDTD, ...) or high-freq. methods (PO, GO, UTD, GTD, ...)
  - 대부분의 실제 문제들은 numerical methods 활용



MoM (Moment method)



FDTD (Finite-Difference Time-Domain)

# 전류분포 모델링: 극소다이폴 모델링

## 전류분포와 Free-space Green's function

- 고주파수 대역 활용 → electrical dimension 증가
- 회로 및 소자의 복잡도 증가 → finer segmentation 필요
- 기술 발전 → numerical method의 계산 복잡도 증가
- 따라서, 기존 *numerical method*의 대안 필요

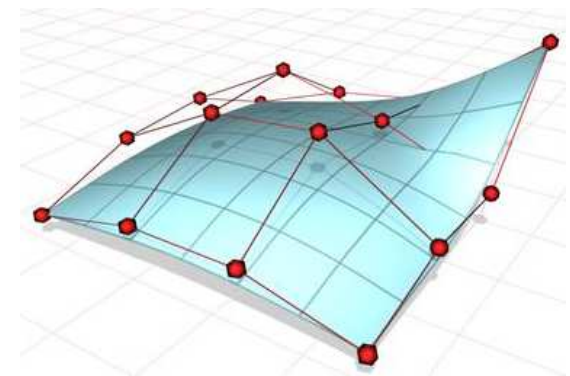
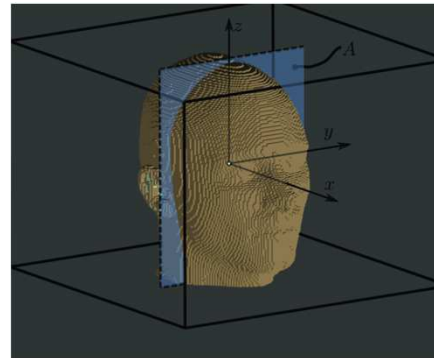
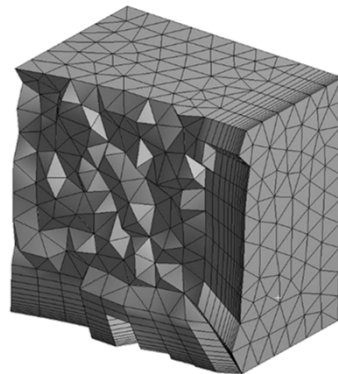
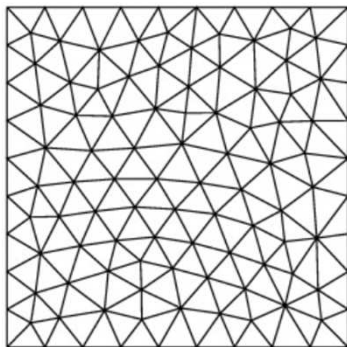
NURBS\*: Non-Uniform Rational Basis Spline

## 전류분포와 Free-space Green's function

- numerical method를 통한 전류분포의 exact solution?

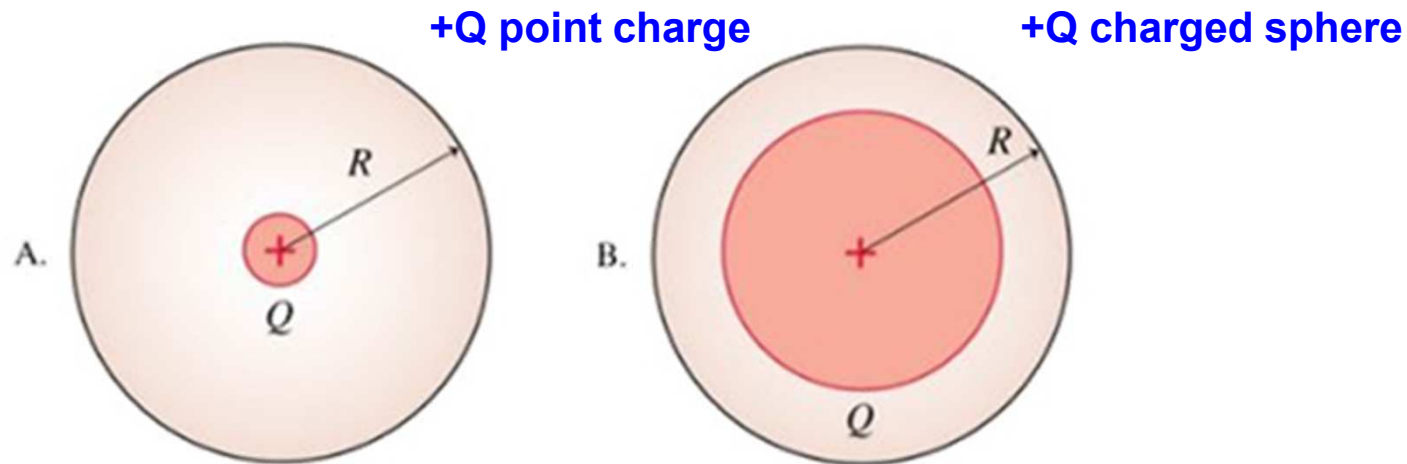
Numerical method	Segmentation model
MoM	triangular or tetrahedron mesh
FDTD	rectangular voxel
PO, PTD	NURBS*

- 결국 solution은, *'induced current model'*



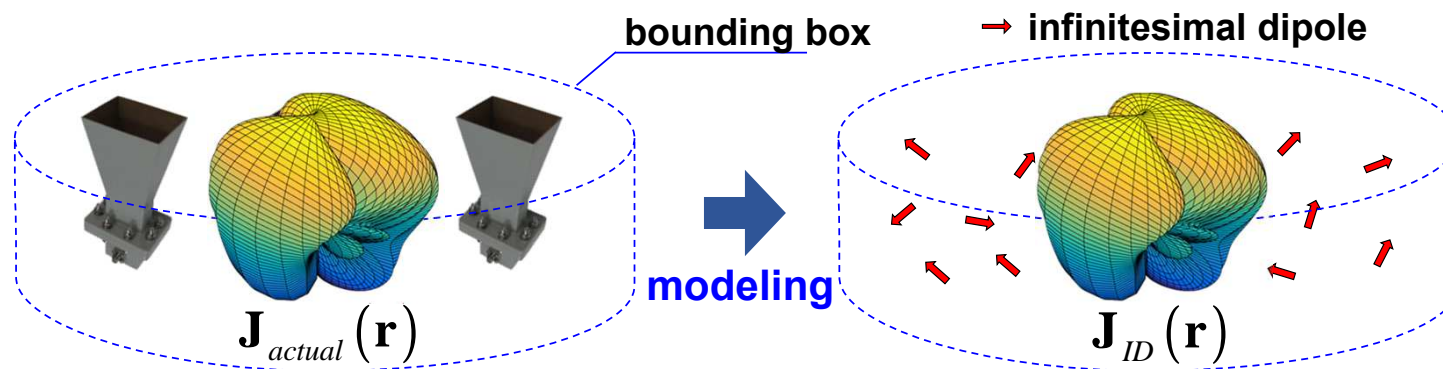
## 전류분포와 Free-space Green's function

- Ex) Gauss' law
  - same field distribution outside the boundary
  - **same** radiated field → **different** current model (O)
  - **same** current model → **different** radiated field (X)



## Basic concept of infinitesimal dipole modeling (IDM)

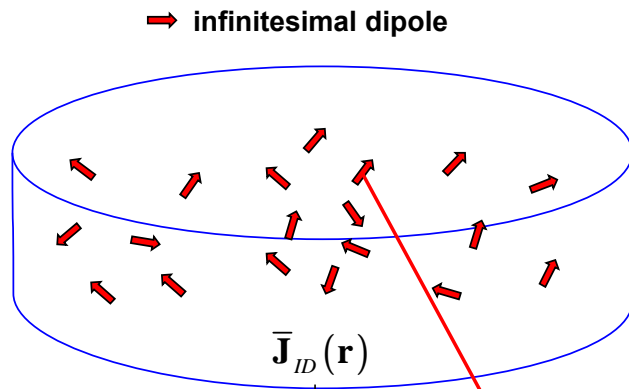
- $\mathbf{E}^{\text{inc}}$  (incident field)  $\rightarrow$   **$\mathbf{J}$  (induced current)  $\rightarrow \mathbf{E}, \mathbf{H}$  (radiated field)**
- radiated field로부터 induced current를 역추적
- 일종의 reverse engineering
- 최적화 된 IDM = an induced current model





## Basic concept of infinitesimal dipole modeling (IDM)

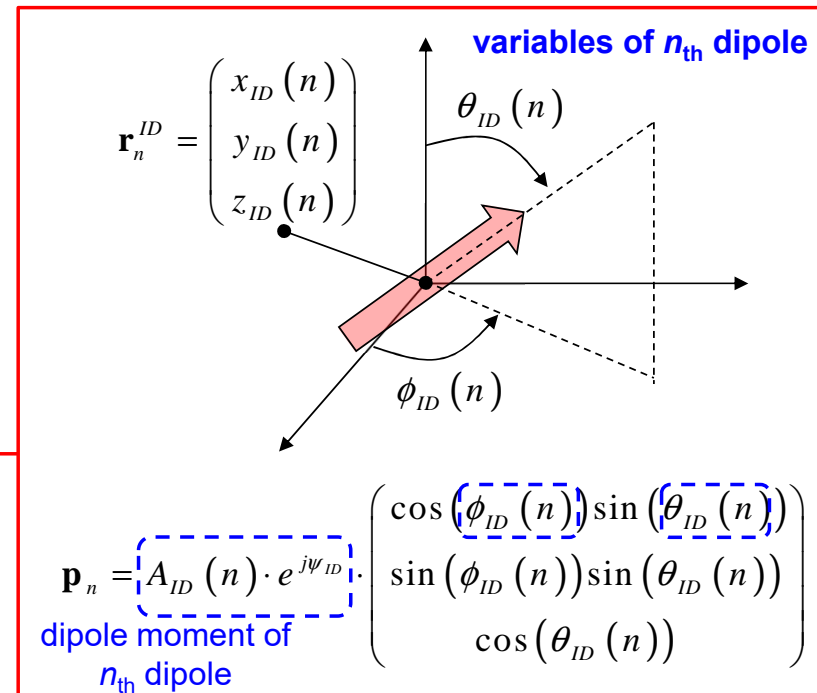
- variables of infinitesimal dipoles
  - position (3) + excitation coefficients (2) + orientation (2) = 7 variables



$$\bar{\mathbf{J}}_{ID}(\mathbf{r})$$

$$\mathbf{J}_{ID}(\mathbf{r}) = \sum_{n=1}^N \mathbf{p}_n \delta(\mathbf{r} - \mathbf{r}_n^{ID})$$

a set of infinitesimal dipoles  
(a set of point current sources)



## Basic concept of infinitesimal dipole modeling (IDM)

- near and far-field of IDM

- near-field

$$\mathbf{E}(\mathbf{r}) = \sum_{n=1}^N \frac{1}{4\pi\omega\epsilon_0} \left[ \boxed{k^2} \left( (\mathbf{n}_n \times \boxed{\mathbf{p}_n}) \times \mathbf{n}_n \right) \frac{e^{-jk r_n}}{\boxed{r_n}} + \left( 3\mathbf{n}_n (\mathbf{n}_n \cdot \mathbf{p}_n) - \mathbf{p}_n \right) \left( \frac{jk}{r_n^2} + \frac{1}{r_n^3} \right) e^{-jk r_n} \right]$$

$\boxed{\mathbf{n}_n} = \left( \frac{x - x_{ID}(n)}{r_n}, \frac{y - y_{ID}(n)}{r_n}, \frac{z - z_{ID}(n)}{r_n} \right)^T = \frac{\mathbf{r} - \mathbf{r}_n^{ID}}{|\mathbf{r} - \mathbf{r}_n^{ID}|}$  unit vector from ID to observation point

$\boxed{r_n} = |\mathbf{r} - \mathbf{r}_n^{ID}|$  distance between ID and observation point

0

- far-field

- using far-field approximation

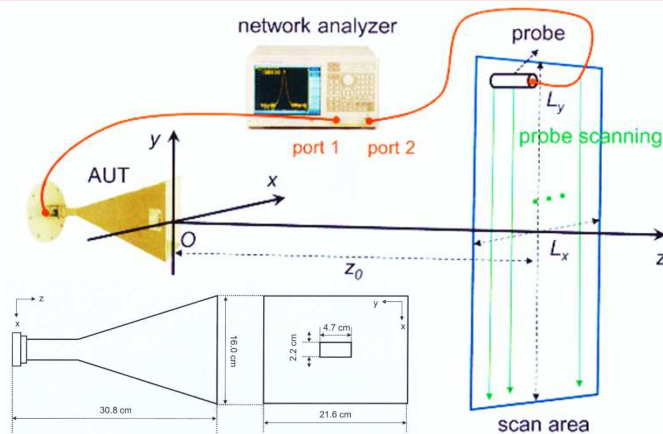
$$\mathbf{E}(\mathbf{r}) = \sum_{n=1}^N \frac{1}{4\pi\omega\epsilon_0} \left[ k^2 \left( \boxed{\hat{\mathbf{r}} \times \mathbf{p}_n} \right) \times \hat{\mathbf{r}} \right] \frac{e^{jk \mathbf{r}_n^{ID} \cdot \hat{\mathbf{r}}}}{r} e^{-jkr}$$

$\mathbf{n}_n \rightarrow \hat{\mathbf{r}}$  ① path differences (positions)

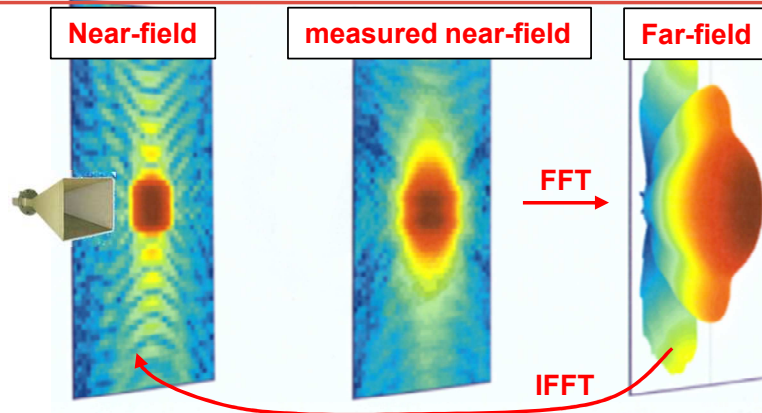
② excitation coefficients and polarizations

## Infinitesimal dipole modeling vs. Fourier transform

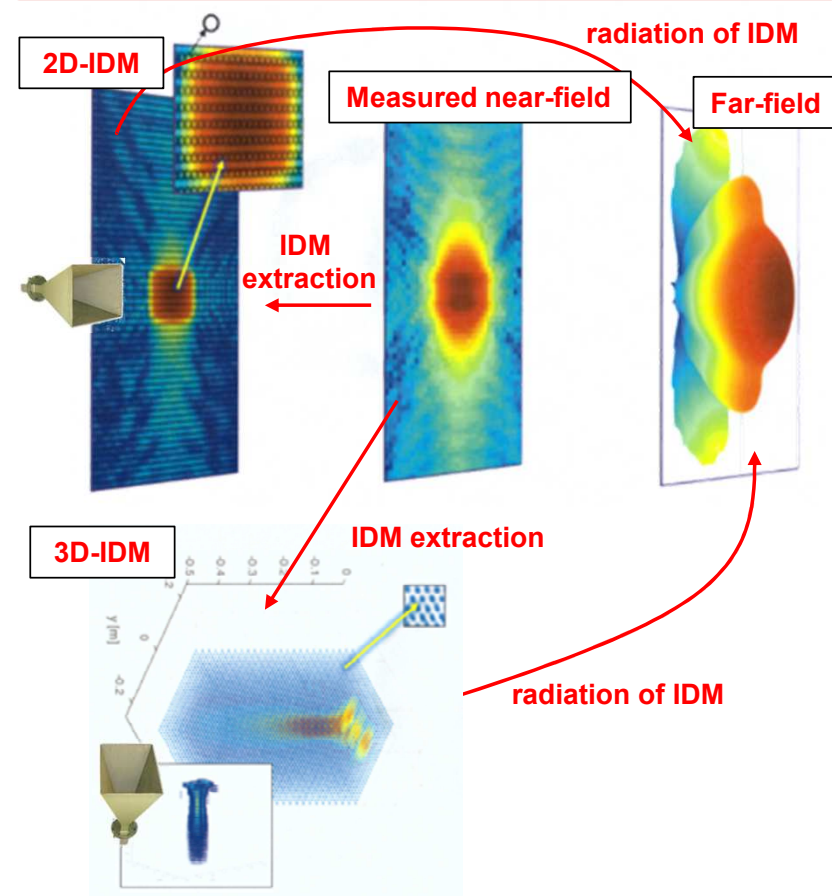
Near-field measurement of Horn antenna



FFT based field transform



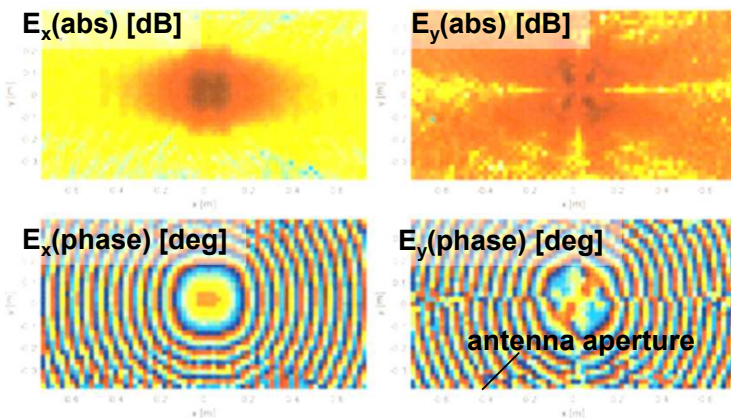
IDM based field transform



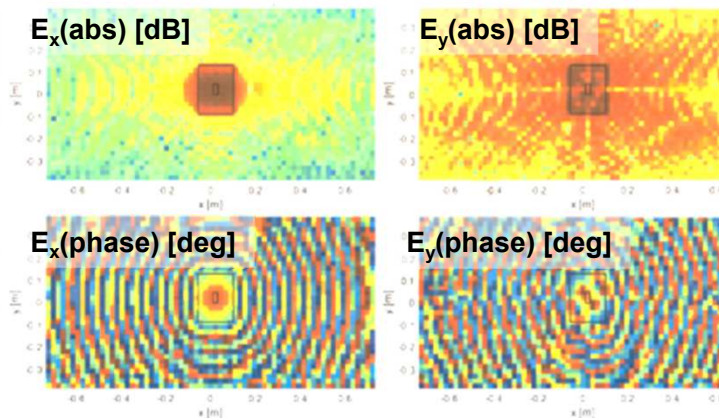
## Infinitesimal dipole modeling vs. Fourier transform

- 결과적인 물리량의 차이
- Fourier transform-based projection → field transform (signal processing)
- IDM-based projection → **induced current model (electromagnetics)**

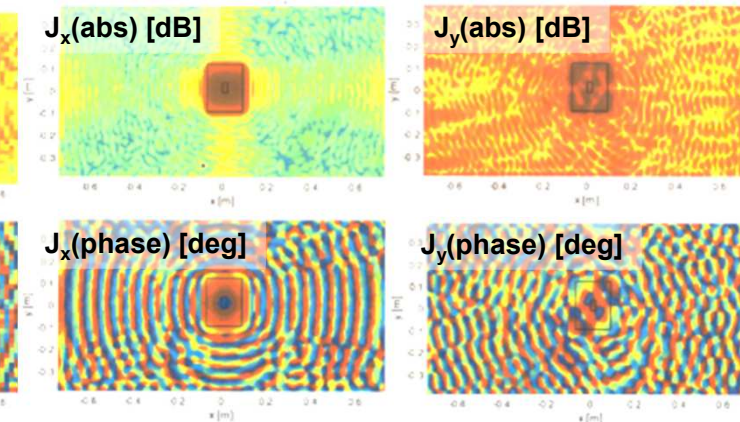
Measured near-field



Near-field (Back-projection, IFFT)



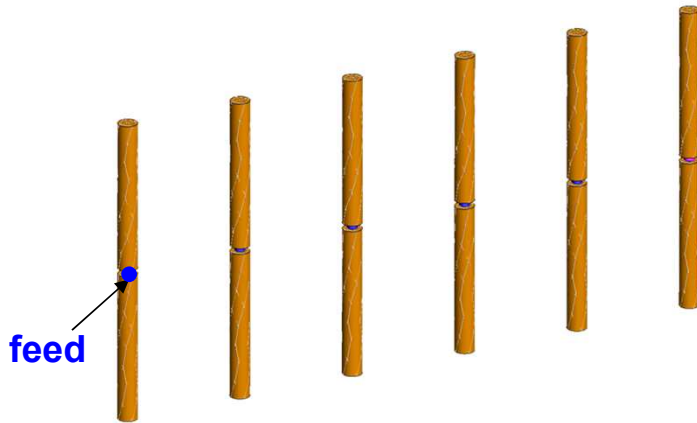
Current model (2D-IDM)





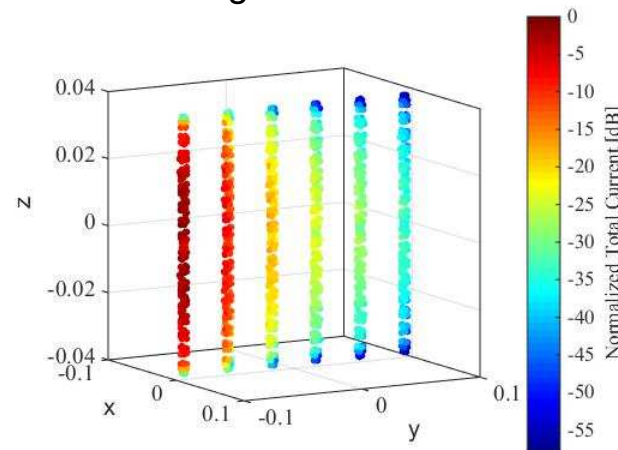
## Verification of infinitesimal dipole modeling (IDM)

- **visual** similarity of current distribution
- MoM vs. IDM
- Ex 1) dipole array antenna



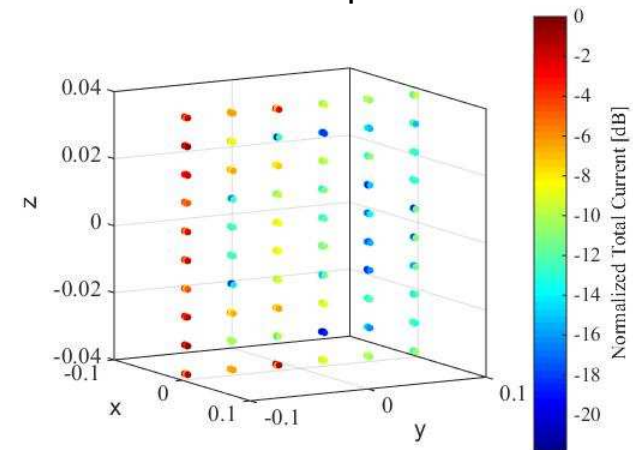
dipole array antenna

# of triangular mesh = 960



current model (MoM)

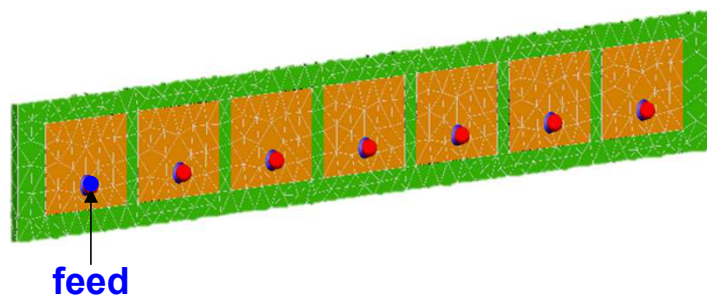
# of infinitesimal dipoles = 240



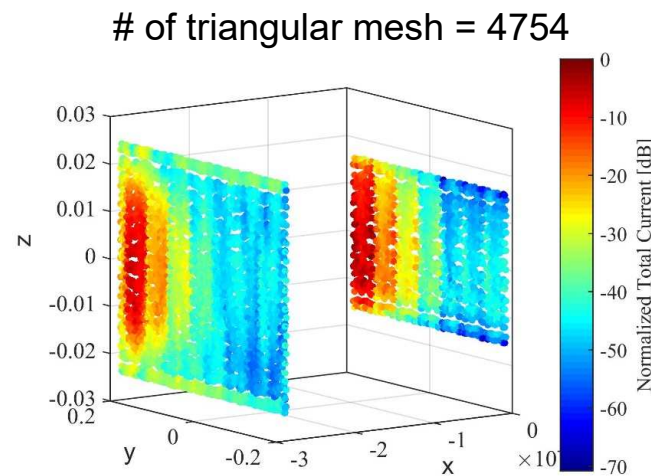
current model (IDM)

## Verification of infinitesimal dipole modeling (IDM)

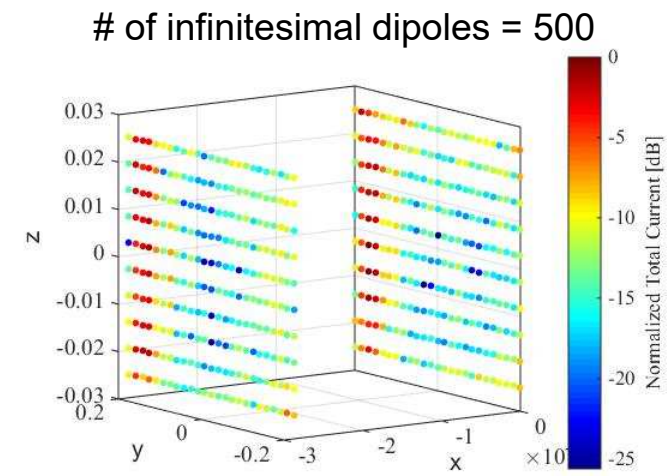
- **visual** similarity of current distribution
- MoM vs. IDM
- Ex 2) patch array antenna



patch array antenna



current model (MoM)

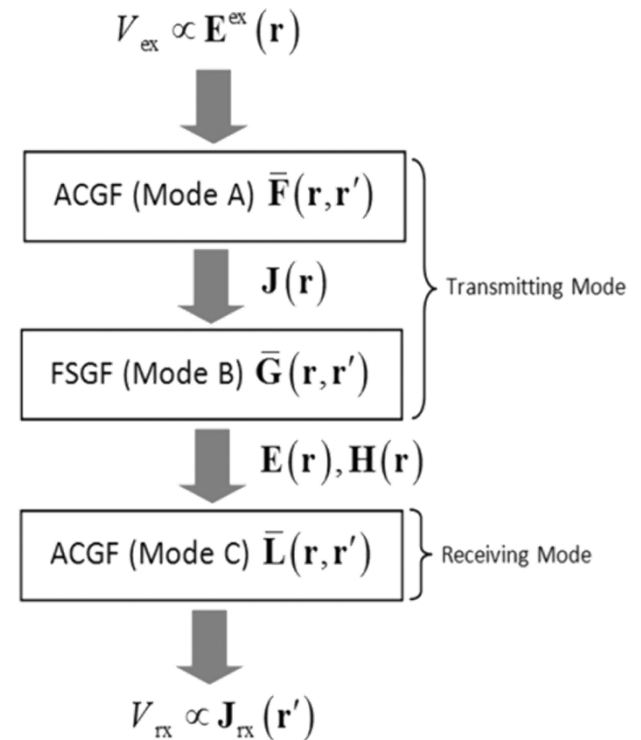
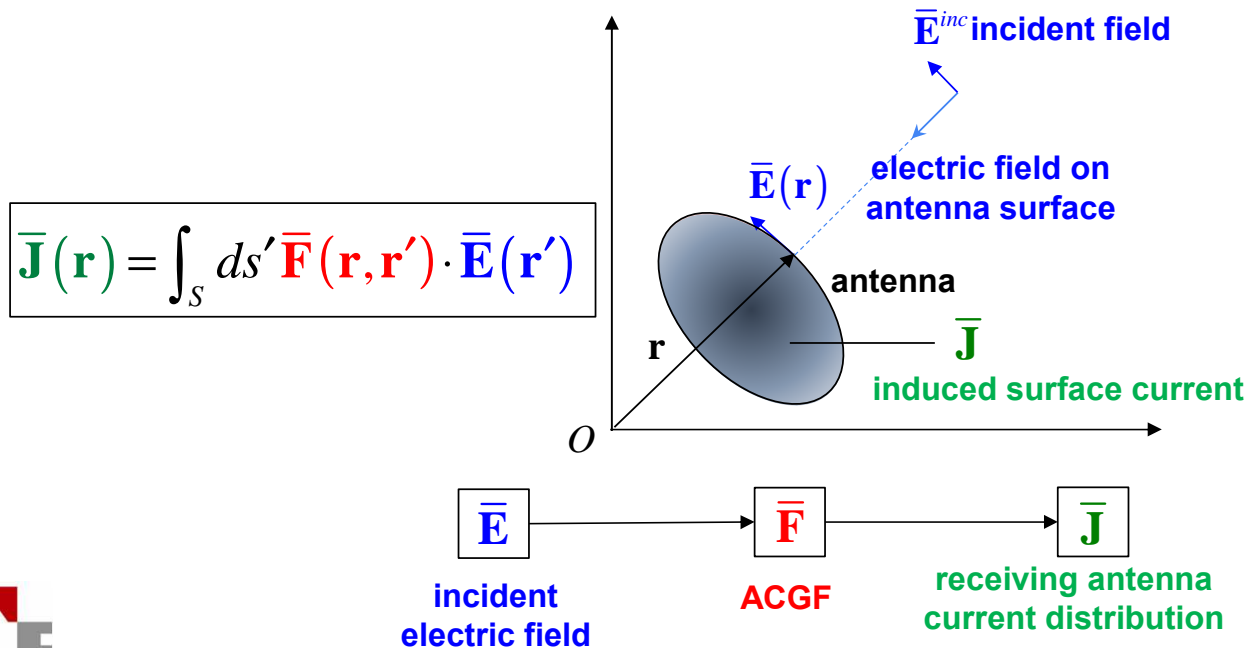


current model (IDM)

## Antenna Current Green's Function (ACGF)

: impulse response represented as **antenna current distribution** with **impulse electric field excitation**

receiving mode analysis using ACGF



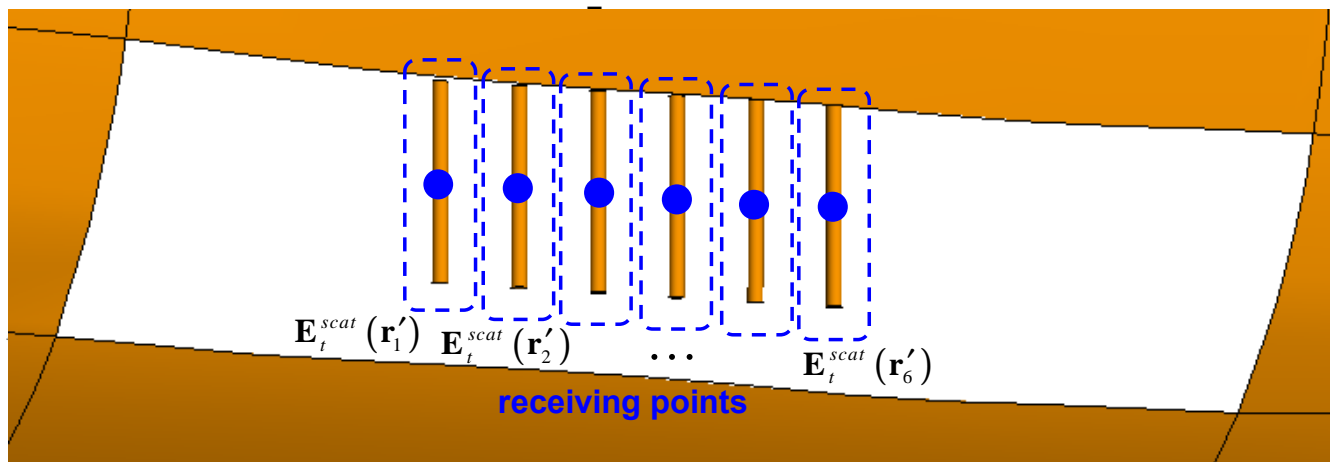
# 극소다이폴 모델의 검증

## IDM은 ACGF로써 활용될 수 있을까?

- 산란파 수신 환경에서의 능동소자패턴 계산

$$v(\mathbf{r}'_m) \square - \frac{Z_L Z_{ANT}}{Z_L + Z_{ANT}} \left\{ \sum_{n=1}^N \sum_{k=1}^K \bar{\mathbf{F}}_{nm}^a(\mathbf{r}_k, \mathbf{r}'_m) \cdot \boxed{\mathbf{E}_t^{dir}(\mathbf{r}_k)} ds_k + \sum_{n=1}^N \bar{\mathbf{F}}_{nm}^a(\mathbf{r}'_n, \mathbf{r}'_m) \cdot \boxed{\mathbf{E}_t^{scat}(\mathbf{r}'_n)} S_n \right\}$$

scattered field  
on receiving point



→ incident direction  
→ polarized direction

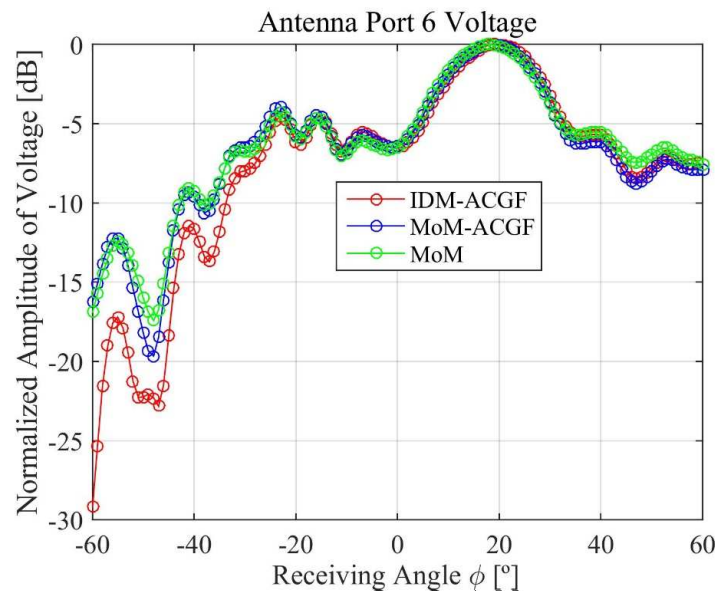


## IDM은 ACGF로써 활용될 수 있을까?

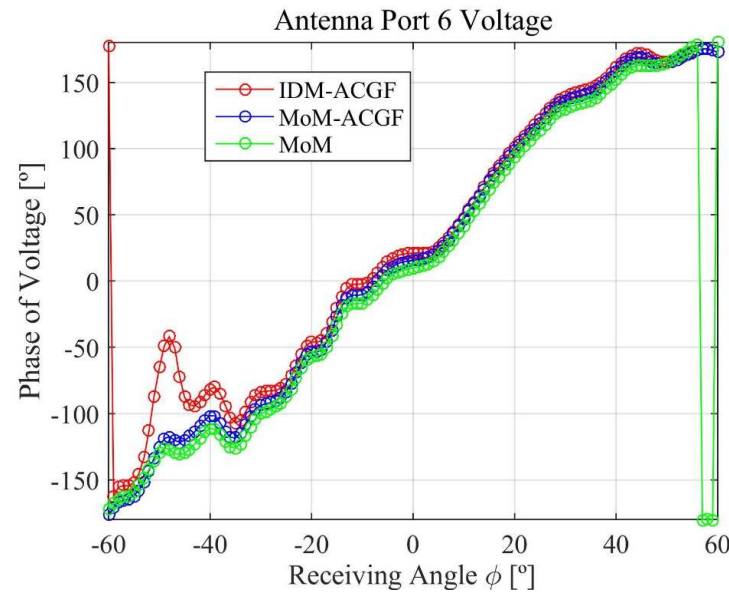
- 산란파 수신 환경에서의 능동소자패턴 계산

$$v(\mathbf{r}'_m) \approx -\frac{Z_L Z_{ANT}}{Z_L + Z_{ANT}} \left\{ \sum_{n=1}^N \sum_{k=1}^K \bar{\mathbf{F}}_{nm}^a(\mathbf{r}_k, \mathbf{r}'_m) \cdot \boxed{\mathbf{E}_t^{dir}(\mathbf{r}_k)} ds_k + \sum_{n=1}^N \bar{\mathbf{F}}_{nm}^a(\mathbf{r}'_n, \mathbf{r}'_m) \cdot \boxed{\mathbf{E}_t^{scat}(\mathbf{r}'_n)} S_n \right\}$$

scattered field  
on receiving point



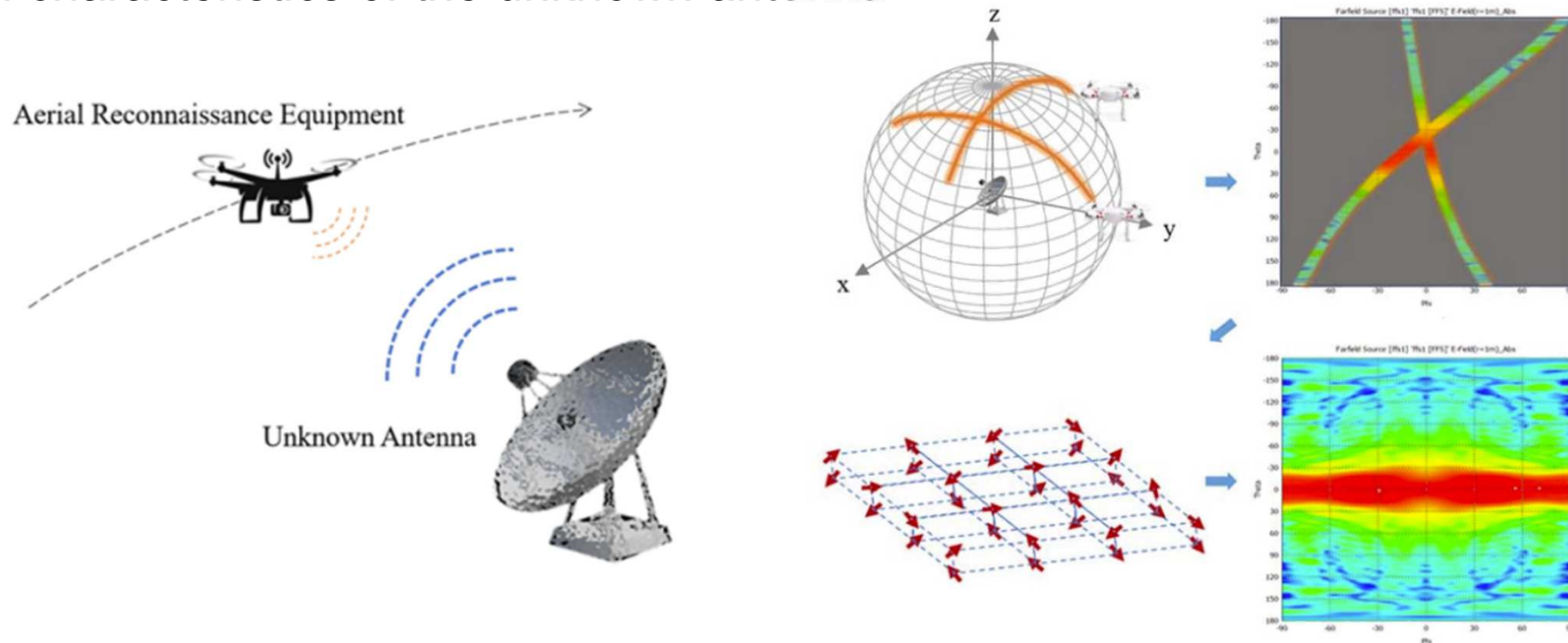
calculated & measured AEP (amplitude)  
( $\phi = 105^\circ, \theta = -60^\circ : 1^\circ : 60^\circ$ )



calculated & measured AEP (phase)  
( $\phi = 105^\circ, \theta = -60^\circ : 1^\circ : 60^\circ$ )

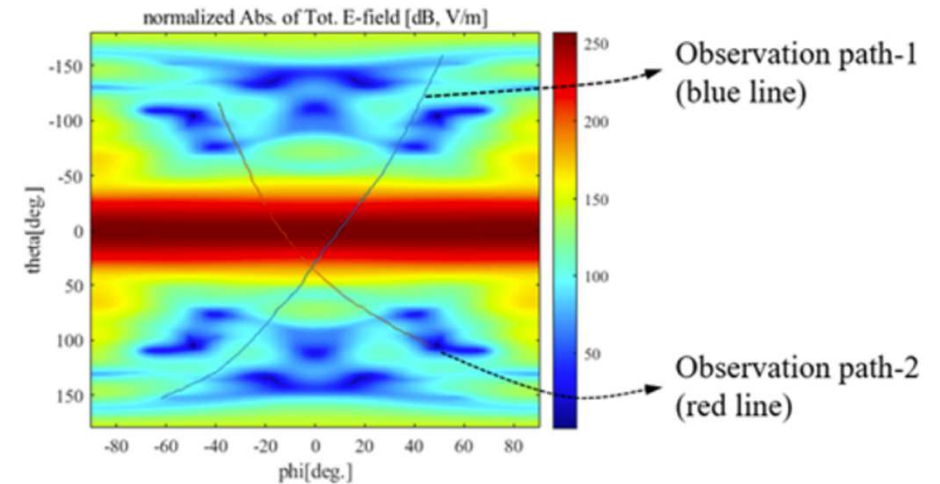
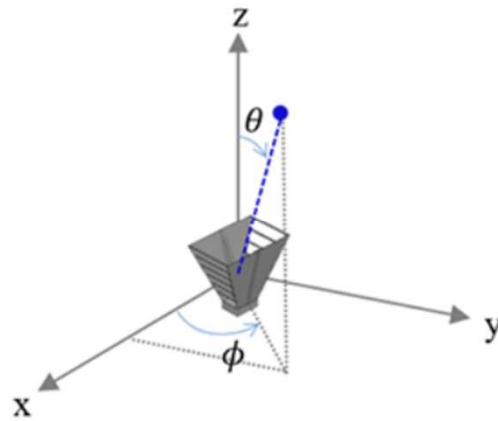
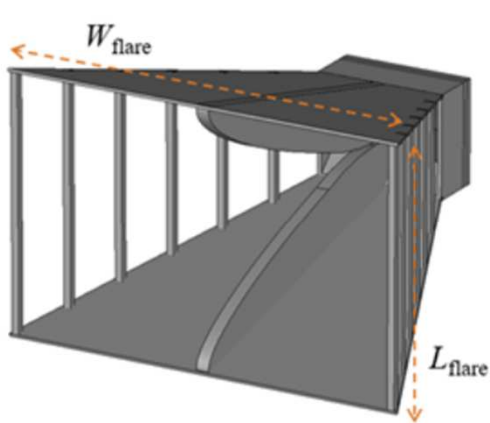
## Transformation of radiated field

- Sparse far-field patterns  $\rightarrow$  Full-coverage of far-field data
- EM characteristics of the unknown antenna



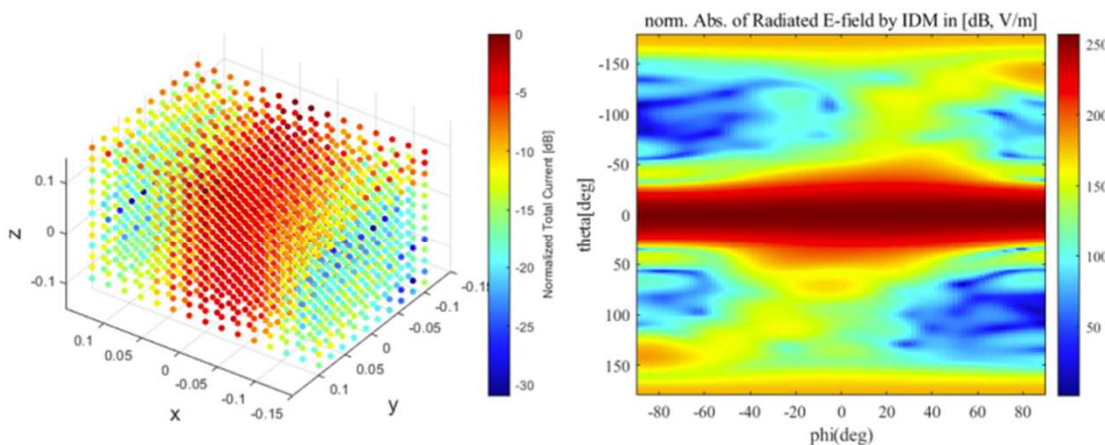
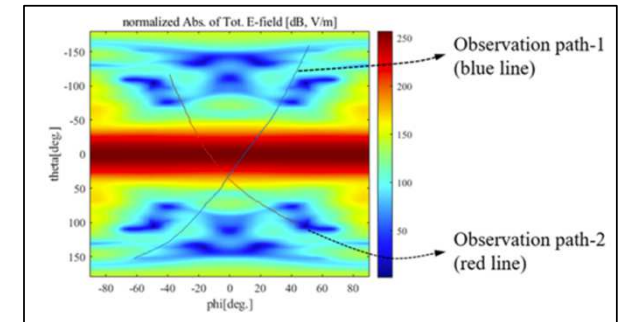
## Transformation of radiated field

- simulation: single-horn antenna (2.4 GHz)
- observed far-field: 1-D angular domain  $\rightarrow$  used to obtain IDM

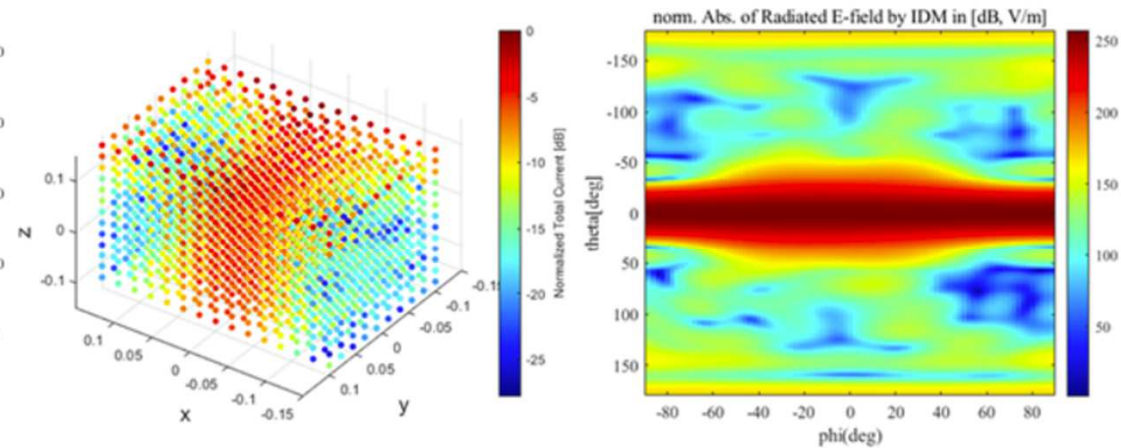


## Transformation of radiated field

- simulation: single-horn antenna (2.4 GHz)
- two different source data  $\rightarrow$  two different results
- “*modeling에 활용된 radiated field의 power portion  $\rightarrow$  modeling accuracy*”



(a)

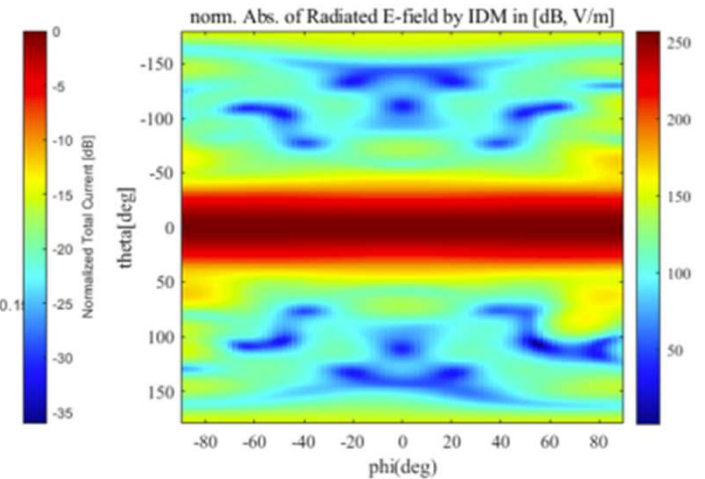
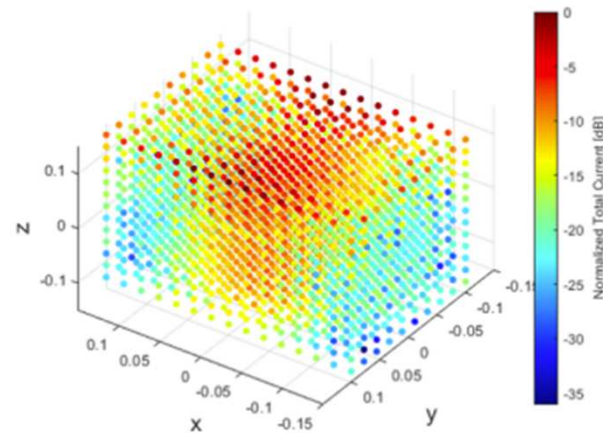
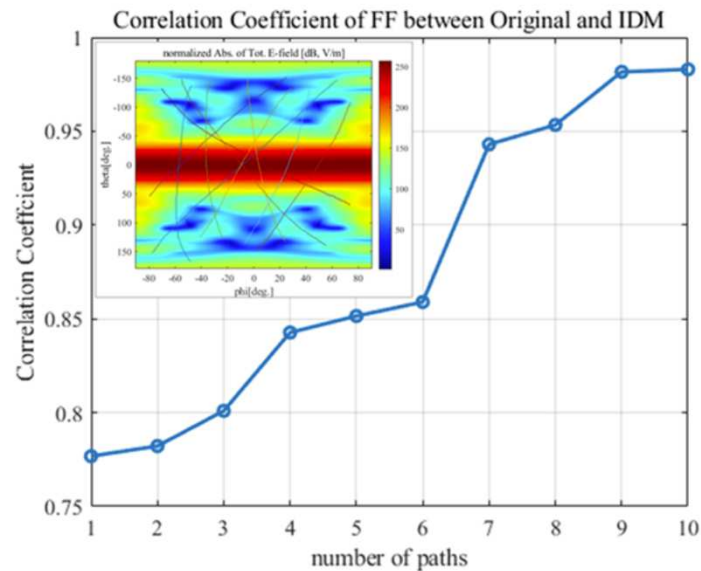


(b)



## Transformation of radiated field

- simulation: single-horn antenna (2.4 GHz)
- *the more source data, the more accurate modeling*



# 극소다이폴 모델의 응용

## IDM 개발 및 발전 방향

- noise와 field accuracy에 대한 sensitivity 정립
- modeling 정확도 개선 (실제 current distribution과의 일치도 향상)
- IDM to current reconstruction / optimization 성능 개선
- 최종 목표: 비접촉 방식의 전류분포 추정